

MIDTERM SOLUTIONS

Wednesday, November 14, 2007, 10:15 - 13:15

Problem 1 [Hypothesis Testing – 20 pts]

(a) The MAP rule for the binary case

$$\frac{p_{Y|X}(y|1)}{p_{Y|X}(y|-1)} \underset{\hat{X}=-1}{\overset{\hat{X}=1}{\leq}} \frac{p_X(-1)}{p_X(1)} = 1. \quad (1)$$

We have,

$$\begin{aligned} \frac{p_{Y|X}(y|1)}{p_{Y|X}(y|-1)} &= \frac{\Pr(Z = W)p(y|1, Z = W) + \Pr(Z = W + 1)p(y|1, Z = W + 1)}{\Pr(Z = W)p(y|-1, Z = W) + \Pr(Z = W + 1)p(y|-1, Z = W + 1)} \\ &= \frac{1/2[p(y|1, Z = W) + p(y|1, Z = W + 1)]}{1/2[p(y|-1, Z = W) + p(y|-1, Z = W + 1)]} \\ &= \frac{f_W(y-1) + f_W(y-2)}{f_W(y+1) + f_W(y)}. \end{aligned} \quad (2)$$

By visual inspection of Figure 1, it is seen that the right-hand side of (2) is > 1 when $y > 1$ (since in this case $f_W(y+1) = 0$ and $f_W(y-1) > f_W(y)$). Similarly, the right-hand side of (2) is < 1 when $y < 0$ (since in this case $f_W(y-2) = 0$ and $f_W(y) > f_W(y-1)$). Therefore it is sufficient to consider $0 \leq y \leq 1$ to find the threshold. In this case we have

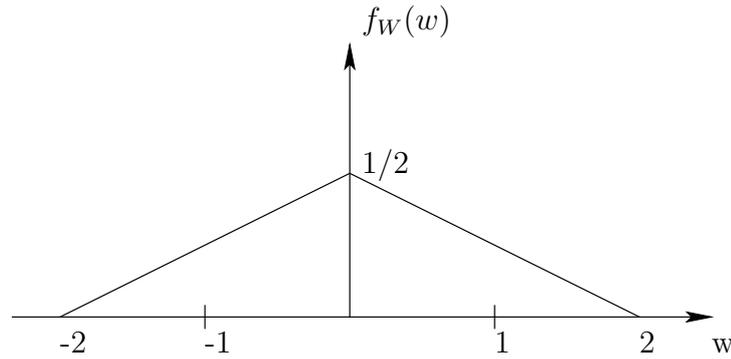


Figure 1:

$$1 = \frac{p_{Y|X}(y|1)}{p_{Y|X}(y|-1)} = \frac{1/2 - |y-1|/4 + 1/2 - |y-2|/4}{1/2 - |y+1|/4 + 1/2 - |y|/4} \quad (3)$$

$$= \frac{1/2 - (1-y)/4 + 1/2 - (2-y)/4}{1/2 - (y+1)/4 + 1/2 - y/4} \quad (4)$$

$$= \frac{1 - (3-2y)/4}{1 - (2y+1)/4}. \quad (5)$$

Solving (5) we get $y = 1/2$.

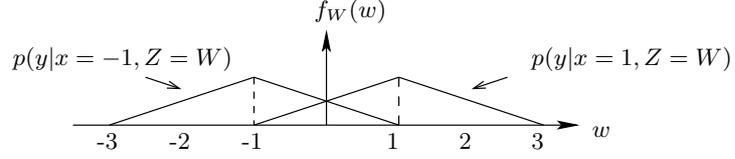


Figure 2:

- (b) First consider the case $Z = W$. By symmetry (see Figure 2), it is easy to argue that in this case the error probability $P_{e,0}(t)$ is symmetric in t , and the threshold minimizing the error probability is $t = 0$. More precisely, $P_{e,0}(t)$ is given by

$$\begin{aligned} P_{e,0}(t) &= \frac{1}{2} \int_t^\infty p(t|x = -1, Z = W) dt + \frac{1}{2} \int_{-\infty}^t p(t|x = 1, Z = W) dt \\ &= \frac{1}{2} \int_t^\infty f_W(t+1) dt + \frac{1}{2} \int_{-\infty}^t f_W(t-1) dt. \end{aligned}$$

Computing this integral for positive values of t we get

$$P_{e,0}(t) = \begin{cases} \frac{1}{16}[(1-t)^2 + (1+t)^2] & \text{if } 0 \leq t < 1 \\ \frac{1}{2}[1 - \frac{(3-t)^2}{8}] & \text{if } 1 \leq t < 3 \\ \frac{1}{2} & \text{if } 3 \leq t \end{cases} \quad (6)$$

(You do not have to compute (6) exactly, as long as you make the following

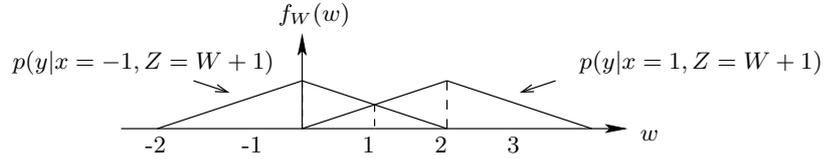


Figure 3:

observations:) By symmetry we have $P_{e,0}(t) = P_{e,0}(-t)$. Also note that $P_{e,0}(t)$ is increasing in $|t|$. Now consider the case $Z = W + 1$. We see in Figure 3 that the symmetry is preserved and the densities are merely shifted right by 1, therefore it is easily argued that $P_{e,1}(t)$ is of the same shape as $P_{e,0}(t)$, symmetric around $t = 1$ and increasing in $|t - 1|$. Therefore the functions $P_{e,0}(t)$ and $P_{e,1}(t)$ take the form in Figure 4. It is seen in this figure that $\max\{P_{e,0}, P_{e,1}\}$ is minimized when $P_{e,0}(t) = P_{e,1}(t)$. Due to symmetry, equality is attained at $t = 1/2$.

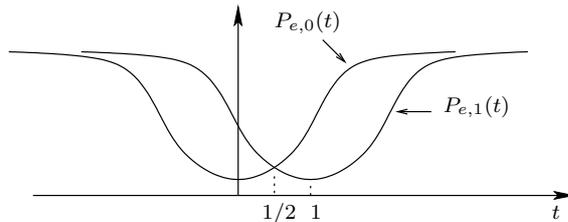


Figure 4:

Problem 2 [Proper Vectors - 20pts]

(a) We can write:

$$h_{>,\mathcal{F}}(f) = \frac{1}{2} + \frac{1}{2}\text{sign}(f)$$

where

$$\text{sign}(f) = \begin{cases} 1 & \text{for } f > 0 \\ 0 & \text{for } f = 0 \\ -1 & \text{for } f < 0 \end{cases}$$

Then, we obtain:

$$\begin{aligned} \hat{x}_{\mathcal{F}}(f) &= \sqrt{2}x_{\mathcal{F}}(f)h_{>,\mathcal{F}}(f) \\ &= \sqrt{2}x_{\mathcal{F}}(f) \left[\frac{1}{2} + \frac{1}{2}\text{sign}(f) \right] \\ &= \frac{x_{\mathcal{F}}(f)}{\sqrt{2}} + \frac{x_{\mathcal{F}}(f)}{\sqrt{2}}\text{sign}(f) \end{aligned}$$

The first term of the last line is symmetric in f (Fourier transform of a real-valued signal is symmetric). So the second term is anti-symmetric, so its inverse Fourier transform is a purely imaginary signal. Hence, by taking the inverse Fourier transform, $\hat{x}(t)$ equals $\frac{x(t)}{\sqrt{2}}$ plus an imaginary term. So $x(t) = \sqrt{2}\text{Re}\{\hat{x}(t)\}$.

(b) $\hat{Z}(f) = \sqrt{2}Z_{\mathcal{F}}(f)h_{>,\mathcal{F}}(f)$ implies that

$$\hat{Z}(t) = \sqrt{2} \int_{-\infty}^{+\infty} h_{>}(\alpha)Z(t - \alpha)d\alpha$$

Then, the pseudocovariance of $\hat{Z}(t)$ is

$$\begin{aligned} \mathbb{E} \left[\hat{Z}(t)\hat{Z}(s) \right] &= \mathbb{E} \left[\sqrt{2} \int_{-\infty}^{+\infty} h_{>}(\alpha)Z(t - \alpha)d\alpha \sqrt{2} \int_{-\infty}^{+\infty} h_{>}(\beta)Z(s - \beta)d\beta \right] \\ &= 2 \int_{-\infty}^{+\infty} \int_{-\infty}^{+\infty} h_{>}(\alpha)h_{>}(\beta)\mathcal{R}_Z(t - \alpha - s + \beta)d\alpha d\beta \\ &= 2 \int_{\alpha} \int_{\beta} h_{>}(\alpha)h_{>}(\beta) \int_{-\infty}^{+\infty} \mathcal{S}_Z(f)e^{j2\pi f(t - \alpha - s + \beta)}df \\ &= 2 \int_f \mathcal{S}_Z(f)e^{j2\pi f(t - s)}h_{>,\mathcal{F}}(f)h_{>,\mathcal{F}}(-f)df \\ &= 0, \end{aligned}$$

since $h_{>,\mathcal{F}}(f)h_{>,\mathcal{F}}(-f) = 0$ for all frequencies except for $f = 0$. $\mathcal{R}_Z(\cdot)$ is the autocorrelation of $Z(t)$ and $\mathcal{S}_Z(f)$ is its Fourier transform. We have also used the fact that $h_{>,\mathcal{F}}(f) = \int_{\alpha} h_{>}(\alpha)e^{-j2\pi f\alpha}d\alpha$.

Hence the integral vanishes. Thus $\hat{Z}(t)$ is proper.

(c)

$$\begin{aligned}\mathbb{E}[Z_E(t)Z_E(s)] &= \mathbb{E}\left[\hat{Z}(t)e^{-j2\pi f_0 t}\hat{Z}(s)e^{-j2\pi f_0 s}\right] \\ &= e^{-j2\pi f_0(t+s)}\mathbb{E}\left[\hat{Z}(t)\hat{Z}(s)\right] \\ &= 0.\end{aligned}$$

(We could have simply argued that $Z_E(t)$ is proper since it is obtained from the proper process $\hat{Z}(t)$ via a linear transformation).

(d) From point (c) we have

$$\begin{aligned}0 = \mathbb{E}[Z_E(t)Z_E(s)] &= \mathbb{E}[\operatorname{Re}\{Z_E(t)\}\operatorname{Re}\{Z_E(s)\} - \operatorname{Im}\{Z_E(t)\}\operatorname{Im}\{Z_E(s)\}] \\ &\quad + j\mathbb{E}[\operatorname{Re}\{Z_E(t)\}\operatorname{Im}\{Z_E(s)\} + \operatorname{Im}\{Z_E(t)\}\operatorname{Re}\{Z_E(s)\}]\end{aligned}$$

implies

$$\mathbb{E}[\operatorname{Re}\{Z_E(t)\}\operatorname{Re}\{Z_E(s)\}] = \mathbb{E}[\operatorname{Im}\{Z_E(t)\}\operatorname{Im}\{Z_E(s)\}]$$

(e) We compute the autocorrelation of $Z_E(t)$:

$$\begin{aligned}\mathbb{E}[Z_E(t)Z_E^*(s)] &= \mathbb{E}[\operatorname{Re}\{Z_E(t)\}\operatorname{Re}\{Z_E(s)\} + \operatorname{Im}\{Z_E(t)\}\operatorname{Im}\{Z_E(s)\}] \\ &\quad - j\mathbb{E}[\operatorname{Re}\{Z_E(t)\}\operatorname{Im}\{Z_E(s)\} - \operatorname{Im}\{Z_E(t)\}\operatorname{Re}\{Z_E(s)\}]\end{aligned}$$

and observe that if the power spectral density of $Z_E(t)$ is symmetric (that is $S_Z(f_0 - f) = S_Z(f_0 + f)$), the autocorrelation of $Z_E(t)$ is real-valued. Thus

$$\mathbb{E}[\operatorname{Re}\{Z_E(t)\}\operatorname{Im}\{Z_E(s)\} - \operatorname{Im}\{Z_E(t)\}\operatorname{Re}\{Z_E(s)\}] = 0$$

On the other hand, from point (d) we have

$$\mathbb{E}[\operatorname{Re}\{Z_E(t)\}\operatorname{Im}\{Z_E(s)\} + \operatorname{Im}\{Z_E(t)\}\operatorname{Re}\{Z_E(s)\}] = 0$$

The last two expressions imply

$$\mathbb{E}[\operatorname{Re}\{Z_E(t)\}\operatorname{Im}\{Z_E(s)\}] = \mathbb{E}[\operatorname{Im}\{Z_E(t)\}\operatorname{Re}\{Z_E(s)\}] = 0$$

which says that the real and imaginary parts of $Z_E(t)$ are uncorrelated. But since they are Gaussian, this implies that they are independent.

Problem 3 [Viterbi Decoder – 20pts]

(a) The MAP rule is given by

$$\begin{aligned}
 & \arg \max_{u_1^n} P_{U_1^n | Y_1^{2n}}(u_1^n | y_1^{2n}) \\
 &= \arg \max_{u_1^n} P_{Y_1^{2n} | U_1^n}(y_1^{2n} | u_1^n) \\
 &= \arg \max_{u_1^n} \prod_{i=1}^n P_{Y_{2i-1}, Y_{2i} | U_i, U_{i-1}}(y_{2i-1}, y_{2i} | u_i, u_{i-1}) \\
 &= \arg \max_{u_1^n} \prod_{i=1}^n P_{Y_{2i-1} | U_i}(y_{2i-1} | u_i) P_{Y_{2i} | U_i, U_{i-1}}(y_{2i} | u_i, u_{i-1}) \\
 &= \arg \max_{u_1^n} \prod_{i=1}^n f(y_{2i-1}, u_i) f(y_{2i}, u_i + u_{i-1})
 \end{aligned}$$

We have used the short notation x_1^n for the vector (x_1, x_2, \dots, x_n) .

(b) Taking the log of the above MAP rule, we get the MAP rule as

$$\arg \max_{u_1^n} \sum_{i=1}^n \log(f(y_{2i-1}, u_i)) + \log(f(y_{2i}, u_i + u_{i-1}))$$

The trellis section for the i th bit is given by

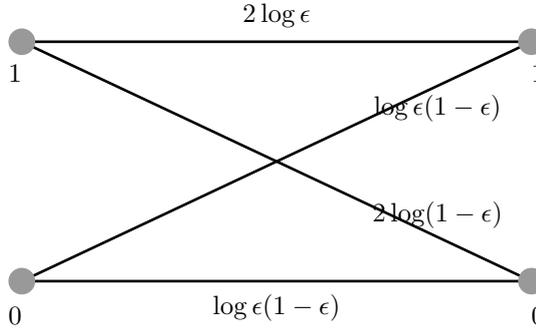


Figure 5: Branch metrics for $y_{2i-1} = 0, y_{2i} = 1$

(c) If the priors are not uniform then the trellis changes to

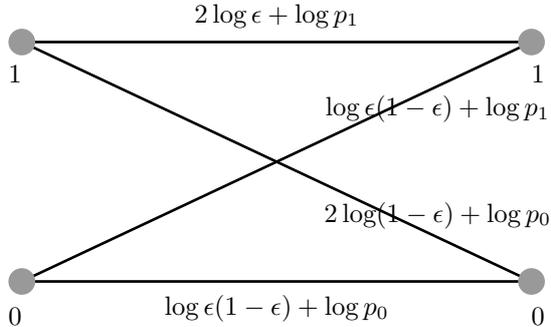


Figure 6: Branch metrics for $y_{2i-1} = 0, y_{2i} = 1$ with unequal priors

Problem 4 [Estimation – 20pts]

Let \tilde{w} be any linear estimator and w_{opt} the linear estimator which satisfies the relation

$$\mathbb{E}[e_{opt}(k)y^*(k-n)] = 0, \text{ for all } n$$

where $e_{opt}(k) = x(k) - \hat{x}_{opt}(k) = x(k) - w_{opt}(k) * y(k) = x(k) - \sum_n w_{opt}(n)y(k-n)$.

The estimator \tilde{w} has an estimation error:

$$\tilde{e}(k) = x(k) - \tilde{w}(k) * y(k) = x(k) - \sum_n \tilde{w}(n)y(k-n).$$

Then

$$\begin{aligned} \mathbb{E}[|\tilde{e}(k)|^2] &= \mathbb{E}[|\tilde{e}(k) - e_{opt}(k) + e_{opt}(k)|^2] \\ &= \mathbb{E}[|\tilde{e}(k) - e_{opt}(k)|^2] + \mathbb{E}[|e_{opt}(k)|^2] + 2\text{Re}\{\mathbb{E}[(\tilde{e}(k) - e_{opt}(k))^* e_{opt}(k)]\} \\ &= \mathbb{E}[|\tilde{e}(k) - e_{opt}(k)|^2] + \mathbb{E}[|e_{opt}(k)|^2] \\ &\quad + 2\text{Re}\left\{\mathbb{E}\left[\left(x(k) - \sum_n \tilde{w}(n)y(k-n) - x(k) + \sum_n w_{opt}(n)y(k-n)\right)^* e_{opt}(k)\right]\right\} \\ &= \mathbb{E}[|\tilde{e}(k) - e_{opt}(k)|^2] + \mathbb{E}[|e_{opt}(k)|^2] \\ &\quad + 2\text{Re}\left\{\sum_n (w_{opt}(n) - \tilde{w}(n))^* \mathbb{E}[e_{opt}(k)y^*(k-n)]\right\} \\ &= \mathbb{E}[|\tilde{e}(k) - e_{opt}(k)|^2] + \mathbb{E}[|e_{opt}(k)|^2] \\ &\geq \mathbb{E}[|e_{opt}(k)|^2]. \end{aligned}$$

where we have used the property of the optimal estimator, that is, $\mathbb{E}[e_{opt}(k)y^*(k-n)] = 0$ for all n .

Therefore we get the “if” part directly due to the inequality $\mathbb{E}[|\tilde{e}(k)|^2] \geq \mathbb{E}[|e_{opt}(k)|^2]$. We get the “only if” part noticing that we need for optimality of *any* other estimator $\mathbb{E}[|\tilde{e}(k) - e_{opt}(k)|^2] = 0$, which means $\tilde{e}(k) = e_{opt}(k)$. This implies that $\sum_n \tilde{w}(n)y(k-n) = \sum_n w_{opt}(n)y(k-n)$, resulting in $\tilde{w}(n) = w_{opt}(n)$ for all n .

Problem 5 [Equalization – 20pts]

(a)

$$\begin{aligned}
 r_{xy}(n) &= \mathbb{E}[x(k)y^*(k-n)] \\
 &= \mathbb{E}\left[x(k)\left(\|p\|\sum_s x^*(s)q^*(k-n-s) + z^*(k-n)\right)\right] \\
 &= \|p\|\sum_s \mathbb{E}[x(k)x^*(s)]q^*(k-n-s) + \mathbb{E}[x(k)z^*(k-n)] \\
 &= \|p\|\mathbb{E}[|x(k)|^2]q^*(-n) \\
 &= \|p\|\mathcal{E}_x q(n)
 \end{aligned}$$

since $\mathbb{E}[x(k)x^*(s)] = \mathcal{E}_x$ for $k = s$ and 0 otherwise.

$$\begin{aligned}
 r_{yy}(n) &= \mathbb{E}[y(k)y^*(k-n)] \\
 &= \mathbb{E}\left[\left(\|p\|\sum_l x(l)q(k-l) + z(k)\right)\left(\|p\|\sum_s x^*(s)q^*(k-n-s) + z^*(k-n)\right)\right] \\
 &= \|p\|^2\sum_l\sum_s \mathbb{E}[x(l)x^*(s)]q(k-l)q^*(k-n-s) + \mathbb{E}[z(k)z^*(k-n)] \\
 &= \|p\|^2\sum_l \mathbb{E}[x(l)x^*(l)]q(k-l)q^*(k-n-l) + q(n)N_0 \\
 &= \|p\|^2\sum_l \mathcal{E}_x q(k-l)q(n-(k-l)) + q(n)N_0 \\
 &= \|p\|^2\sum_l \mathcal{E}_x q(l)q(n-l) + q(n)N_0 \\
 &= \|p\|^2\mathcal{E}_x q(n) * q(n) + q(n)N_0.
 \end{aligned}$$

(b)

$$\begin{aligned}
 S_{xy}(D) &= \mathcal{D}\{r_{xy}(n)\} \\
 &= \mathcal{D}\{\|p\|\mathcal{E}_x q(n)\} \\
 &= \|p\|\mathcal{E}_x Q(D)
 \end{aligned}$$

$$\begin{aligned}
 S_{yy}(D) &= \mathcal{D}\{r_{yy}(n)\} \\
 &= \mathcal{D}\{\|p\|^2\mathcal{E}_x q(n) * q(n) + q(n)N_0\} \\
 &= \|p\|^2\mathcal{E}_x Q(D)Q(D) + Q(D)N_0 \\
 &= \|p\|^2\mathcal{E}_x Q^2(D) + Q(D)N_0
 \end{aligned}$$

(c)

$$\begin{aligned} 0 = \mathbb{E}[e_{opt}(k)y^*(k-n)] &= \mathbb{E}[(x(k) - w_{opt}(k) * y(k))y^*(k-n)] \\ &= \mathbb{E}\left[x(k)y^*(k-n) - \sum_l w_{opt}(l)y(k-l)y^*(k-n)\right] \\ &= \mathbb{E}[x(k)y^*(k-n)] - \sum_l w_{opt}(l)\mathbb{E}[y(k-l)y^*(k-n)] \\ &= r_{xy}(n) - \sum_l w_{opt}(l)r_{yy}(n-l) \\ &= r_{xy}(n) - w_{opt}(n) * r_{yy}(n) \end{aligned}$$

So $r_{xy}(n) = w_{opt}(n) * r_{yy}(n)$. Or in D-domain $S_{xy}(D) = W_{opt}(D)S_{yy}(D)$. The optimal filter $W_{opt}(D)$ is

$$\begin{aligned} W_{opt}(D) &= \frac{S_{xy}(D)}{S_{yy}(D)} \\ &= \frac{\|p\|\mathcal{E}_x Q(D)}{\|p\|^2 \mathcal{E}_x Q^2(D) + Q(D)N_0} \end{aligned}$$